

Linear Systems and Signals

Laplace transform properties, part 1

Anand D. Sarwate

Department of Electrical and Computer Engineering
Rutgers, The State University of New Jersey

2020



Learning objectives

The learning objectives for this section are:

- understand properties of the Laplace transform
- use transform properties to find Laplace transforms

We will use

$$\underline{x(t)} \xleftrightarrow{\mathcal{L}} \underline{X(s)} \quad \text{or} \quad \underline{\mathcal{L}\{x(t)\}} = \underline{X(s)} \quad (1)$$

and label the ROC \mathcal{R} .



The Laplace transform is linear

$$\mathcal{L}\{x(t)\} = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt \quad (2)$$

We already saw that the Laplace transform is linear. If

$$\underline{x_1(t) \xleftrightarrow{\mathcal{L}} X_1(s)} \quad \underline{x_2(t) \xleftrightarrow{\mathcal{L}} X_2(s)} \quad (3)$$

with ROCs \mathcal{R}_1 and \mathcal{R}_2 , then for scalars a_1 and a_2 ,

$$\underline{a_1 x_1(t) + a_2 x_2(t)} \xleftrightarrow{\mathcal{L}} \underline{a_1 X_1(s) + a_2 X_2(s)}. \quad (4)$$

The ROC contains $\mathcal{R}_1 \cap \mathcal{R}_2$.



Time shift

What if we delay $x(t)$ by t_0 ?

$$\underbrace{x(t - t_0)}_{\text{delayed version of } x} \xleftrightarrow{\mathcal{L}} \int_{-\infty}^{\infty} \underbrace{x(t - t_0)}_{\substack{\uparrow \\ \tau = t - t_0}} e^{-st} dt = \int_{-\infty}^{\infty} x(\tau) e^{-s(\tau + t_0)} d\tau \quad (5)$$

$$= \underbrace{e^{-st_0}} \underbrace{X(s)} \quad (6)$$

With the same ROC. It turns out we have the Laplace transform pair

$$\underline{\delta(t) \xleftrightarrow{\mathcal{L}} 1} \quad (7)$$

Which means

LTI system
for pulse response
that delays the input by t_0

$$\underline{\delta(t - t_0) \xleftrightarrow{\mathcal{L}} e^{-st_0}} \quad (8)$$

In words: a system that delays by t_0 has a Laplace transform e^{-st_0} .



Example

Find the Laplace transform of

$$x(t) = 3 \cos(4\pi t)u(t) + e^{-3t}u(t-3) \quad (9)$$

We can use linearity and time shifts:

$$x(t) = 3\mathcal{L}\{\cos(4\pi t)u(t)\} + \mathcal{L}\left\{e^{-9}e^{-3(t-3)}u(t-3)\right\} \quad (10)$$

$$= \frac{3s}{s^2 + (4\pi)^2} + \frac{e^{-9}e^{-3s}}{s+3} \quad (11)$$

The ROC is $\{s : \Re\{s\} > 0\} \cap \{s : \Re\{s\} > -3\} = \{s : \Re\{s\} > 0\}$.



s -domain shift

What about a shift in s instead of a shift in time? Using the same type of argument as time shifts, we get

$$\underline{e^{s_0 t} x(t)} \xleftrightarrow{\mathcal{L}} \underline{X(s - s_0)} \quad (12)$$

Now the ROC shifts by $\Re\{s_0\}$: if $s_0 > 0$ it shifts to the right, and if $s_0 < 0$ it shifts to the left.



Example

Find the Laplace transform of

$$x(t) = e^{-at} \sin(\omega_0 t) u(t). \quad (13)$$

*decaying exp.
s₀ = -a*

We have

$$\sin(\omega_0 t) u(t) \xleftrightarrow{\mathcal{L}} \frac{\omega_0}{s^2 + \omega_0^2}. \quad (14)$$

*shift this
X(s) by -a*

So

$$\underline{e^{-at} \sin(\omega_0 t) u(t)} \xleftrightarrow{\mathcal{L}} \frac{\omega_0}{\underline{(s + a)^2 + \omega_0^2}} \quad (15)$$

*where are
the poles?*

with ROC $\{\Re\{s\} > -a\}$ (shifted to the left by a).



Time dilation/compression

If we look at $x(at)$, we can plug into the transform to get

$$\underline{x(at)} \xleftrightarrow{\mathcal{L}} \int_{-\infty}^{\infty} \underline{x(at)} e^{-st} dt = \int_{-\infty}^{\infty} \frac{1}{|a|} x(\tau) e^{-(s/a)\tau} d\tau \quad (16)$$

$$= \frac{1}{|a|} X\left(\frac{s}{a}\right) \quad (17)$$

The ROC is:

$$\mathcal{R}(x(at)) = \{s : s/a \in \mathcal{R}\}. \quad (18)$$

If $|a| < 1$ we are slowing down $\underline{x(t)}$ and the ROC shrinks by a factor of a and if $|a| > 1$ we are speeding up $\underline{x(t)}$ and the ROC expands by a factor of a .



Conjugation

If $x(t) \xleftrightarrow{\mathcal{L}} X(s)$ then

$$\underline{x^*(t)} \xleftrightarrow{\mathcal{L}} = \int_{-\infty}^{\infty} \underline{x^*(t)} e^{-st} dt = \left(\int_{-\infty}^{\infty} \underline{x(t)} e^{-s^*t} dt \right)^* \quad (19)$$

$$= X^*(s^*). \quad \text{X}(s^*) \quad (20)$$

Why is this useful? If $x(t)$ is real, then $\underline{x(t) = x^*(t)}$ and $\underline{X(s) = X^*(s^*)}$. That means if $X(s)$ has a pole or zero at $\underline{s = s_0}$, then it also has one at $\underline{s = s_0^*}$.

We say that for real signals, the poles and zeros appear in conjugate pairs.



Try it yourself

Problem

Try finding the following Laplace transforms using the properties:

$$x(t) = \delta(t - 2) - 3e^{-5t}u(t) \quad (21)$$

$$x(t) = e^{-3t}u(t) + e^{-4t}\cos(6t)u(t) \quad (22)$$

$$x(t) = e^{-2|t|} \quad (23)$$

