

Linear Systems and Signals

Laplace transform examples

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2020



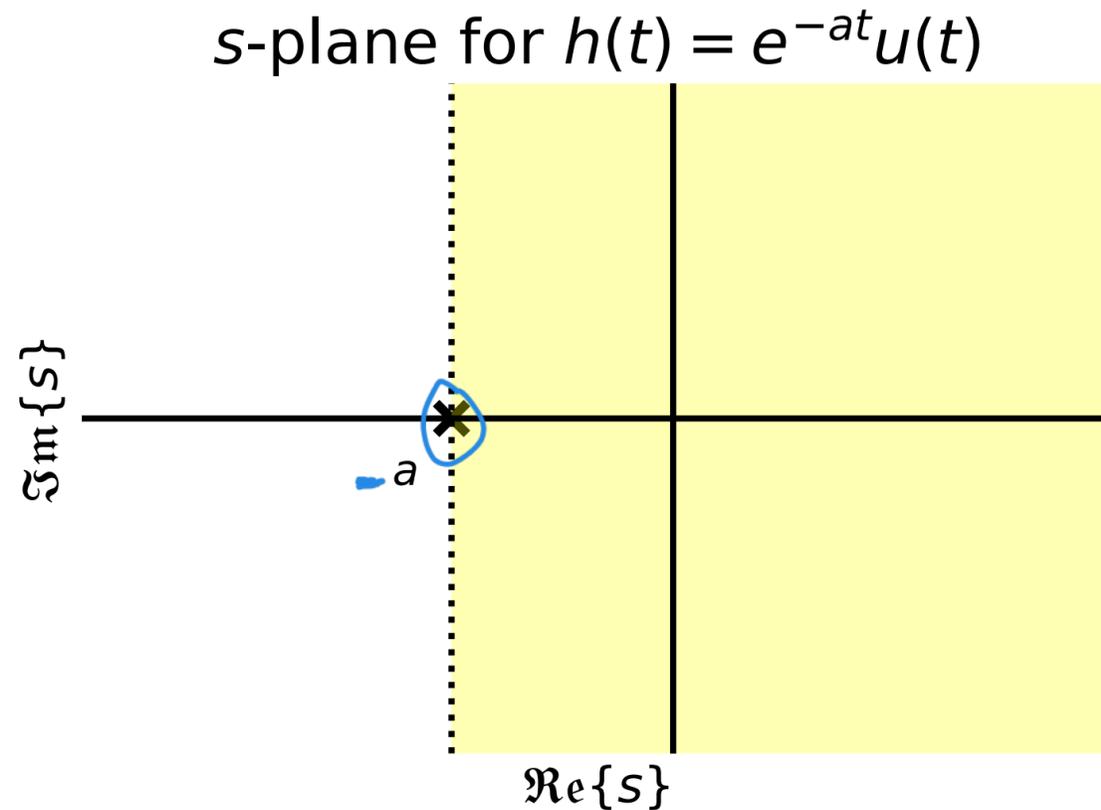
Learning objectives

The learning objectives for this section are:

- use the definition to compute Laplace transforms
- identify the region of convergence for a Laplace transform



General exponentials



If $h(t) = e^{-at}u(t)$ we get a general formula:

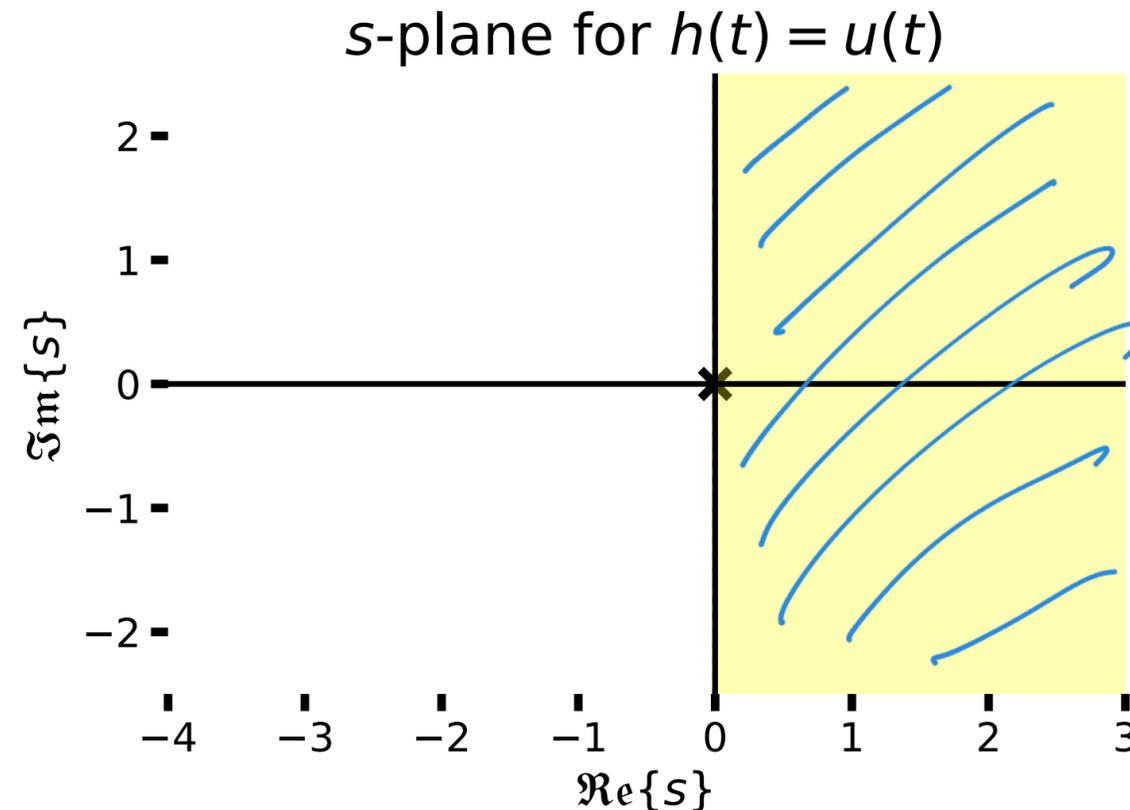
$$H(s) = \int_{-\infty}^{\infty} \underbrace{e^{-at}}_{\text{blue underline}} \underbrace{u(t)}_{\text{blue underline}} e^{-st} dt = \int_0^{\infty} e^{-(s+a)t} dt = \frac{1}{s+a}. \quad (1)$$

$\Re\{s\} > -a$

What about the ROC? We need $\Re\{s\} > -a$. This has a *pole* at $-a$.



The unit step



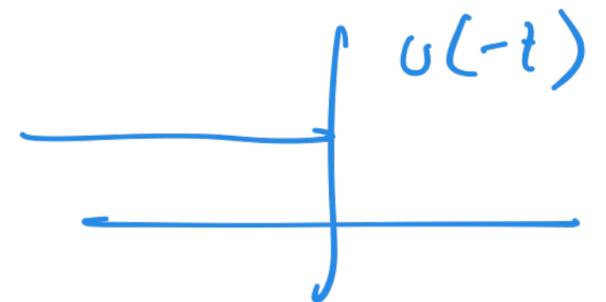
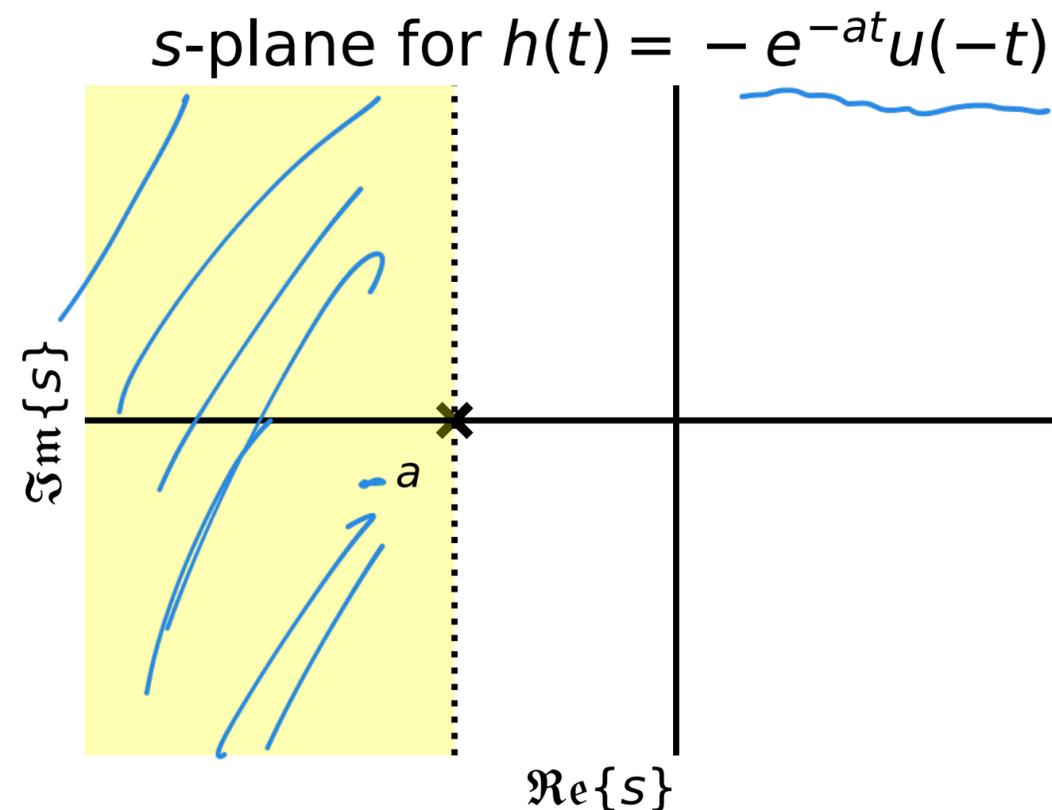
If $h(t) = u(t)$ then the system is an integrator:

$$H(s) = \int_{-\infty}^{\infty} \underbrace{u(t)} e^{-st} dt = \int_0^{\infty} e^{-st} dt = \frac{1}{s}. \quad (2)$$

What about the ROC? We need $\Re\{s\} > 0$. We can also see this as the previous example with $a = 0$.



Anticausal systems



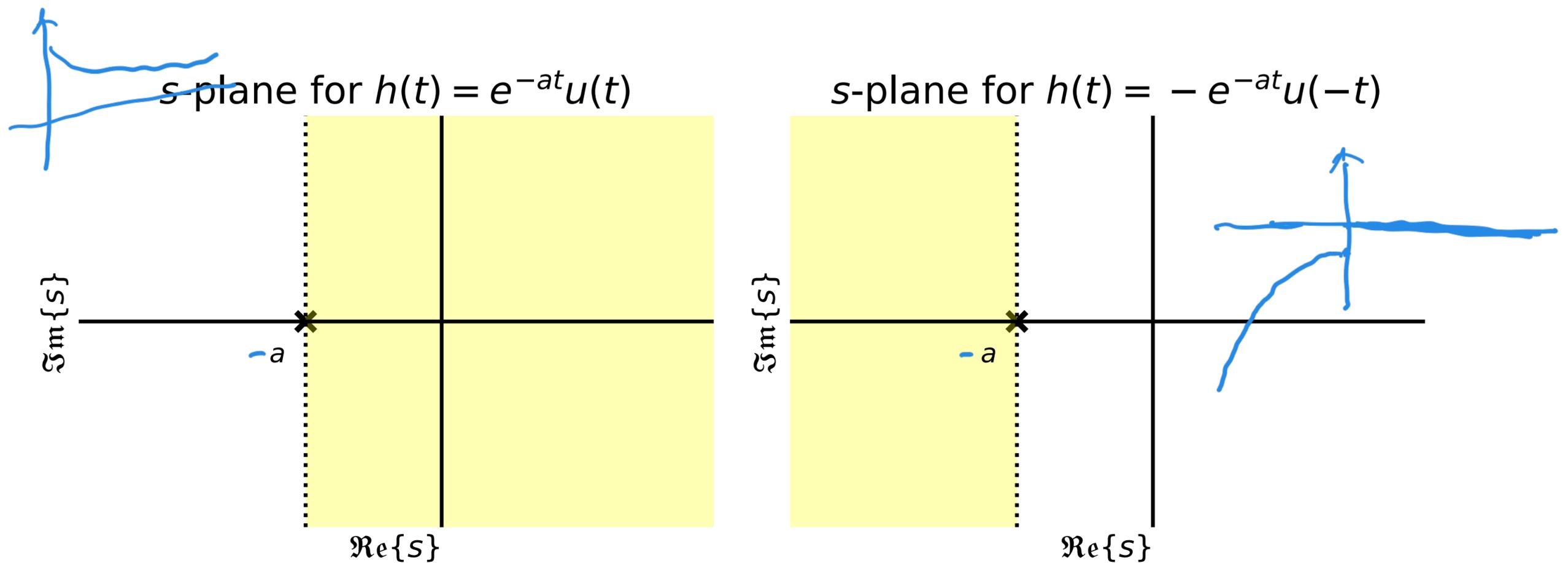
If $h(t) = -e^{-at}u(-t)$ we get a general formula:

$$H(s) = \int_{-\infty}^{\infty} -e^{-at}u(-t)e^{-st} dt = \int_{-\infty}^{\infty} -e^{-(s+a)t} dt = \frac{1}{s+a} \quad (3)$$

Handwritten notes: $s+a < 0$ (with an arrow pointing to the exponent), and a blue circle around the integration limits $-\infty$ to ∞ in the integral.

What about the ROC? We need $\Re\{s\} < -a$. This has a *pole* at $-a$ but the ROC goes the other way.

Two signals with the same transform?



We have both

$$e^{-at}u(t) \xleftrightarrow{\mathcal{L}} \frac{1}{s+a} \qquad -e^{-at}u(-t) \xleftrightarrow{\mathcal{L}} \frac{1}{s+a} \quad (4)$$

The difference is that their ROCs are different. To write a Laplace transform, you always have to specify the ROC!

A cosine

Let's take $h(t) = \cos(\omega t)u(t)$. First step is to Eulerize:

$$\cos(\omega t)u(t) = \left(\frac{1}{2}e^{j\omega t} + \frac{1}{2}e^{-j\omega t} \right) u(t). \quad (5)$$

Taking the Laplace transform:

$$H(s) = \int_{-\infty}^{\infty} \left(\frac{1}{2}e^{j\omega t}u(t) + \frac{1}{2}e^{-j\omega t} \right) u(t)e^{-st} dt \quad (6)$$

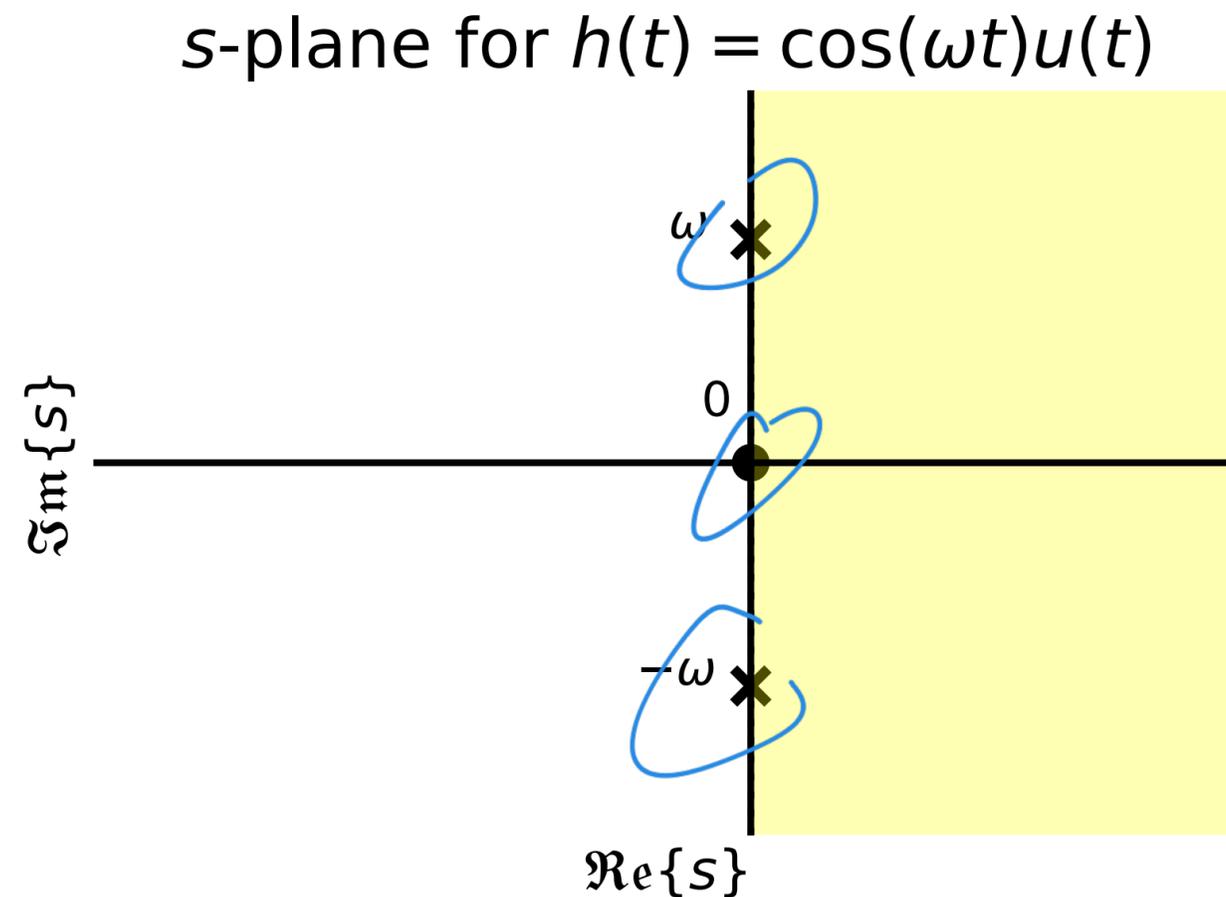
$$= \frac{1}{2} \int_0^{\infty} e^{-(s-j\omega)t} dt + \frac{1}{2} \int_0^{\infty} e^{-(s+j\omega)t} dt \quad (7)$$

$$= \frac{1/2}{s-j\omega} + \frac{1/2}{s+j\omega} \quad (8)$$

$$= \frac{s}{s^2 + \omega^2} \quad \text{degree 2} : 2 \text{ roots} \Rightarrow 2 \text{ poles} \quad (9)$$



A cosine



*pole-zero
diagram*

So we have

$$\cos(\omega t)u(t) \xleftrightarrow{\mathcal{L}} \frac{s}{s^2 + \omega^2}$$

(10)

This is zero at $s = 0$ and blows up (has poles) at $s = \pm j\omega$



Try it yourself

Problem

Try to find the following Laplace transforms from the definition, without using the Laplace Transform table. However, in practice you can use the table to just look up the relevant transforms.

- $e^{-2t}u(t) - e^{3t}u(-t)$
- $e^{-t} \cos(2\pi t)u(t)$
- $\sin(\omega t)u(t)$

Don't forget the region of convergence!

