

# Linear Systems and Signals

## Laplace transform properties, part 1

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2020



# Learning objectives

The learning objectives for this section are:

- understand properties of the Laplace transform
- use transform properties to find Laplace transforms

We will use

$$\underline{x(t)} \stackrel{\mathcal{L}}{\longleftrightarrow} \underline{X(s)} \quad \text{or} \quad \underline{\mathcal{L}\{x(t)\}} = \underline{X(s)} \quad (1)$$

and label the ROC  $\mathcal{R}$ .



# The Laplace transform is linear

$$\mathcal{L}\{x(t)\} = X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt \quad (2)$$

We already saw that the Laplace transform is linear. If

$$\underline{x_1(t)} \xleftrightarrow{\mathcal{L}} \underline{X_1(s)} \quad \underline{x_2(t)} \xleftrightarrow{\mathcal{L}} \underline{X_2(s)} \quad (3)$$

with ROCs  $\underline{\mathcal{R}_1}$  and  $\underline{\mathcal{R}_2}$ , then for scalars  $\underline{a_1}$  and  $\underline{a_2}$ ,

$$\underline{a_1 x_1(t) + a_2 x_2(t)} \xleftrightarrow{\mathcal{L}} \underline{a_1 X_1(s) + a_2 X_2(s)}. \quad (4)$$

The ROC contains  $\underline{\mathcal{R}_1 \cap \mathcal{R}_2}$ .



# Time shift

What if we delay  $x(t)$  by  $t_0$ ?

$$x(t - t_0) \xleftrightarrow{\mathcal{L}} \int_{-\infty}^{\infty} x(t - t_0) e^{-st} dt = \int_{-\infty}^{\infty} x(\tau) e^{-s(\tau + t_0)} d\tau \quad (5)$$

$\tau = t - t_0$

$$= e^{-st_0} X(s) \quad (6)$$

*Handwritten notes:*  
 - Under  $x(t - t_0)$ : delayed version of  $x$   
 - Under  $e^{-st}$ :  $\tau = t - t_0$   
 - Under  $e^{-s(\tau + t_0)}$ :  $\tau = t - t_0$

With the same ROC. It turns out we have the Laplace transform pair

$$\delta(t) \xleftrightarrow{\mathcal{L}} 1 \quad (7)$$

Which means

$$\delta(t - t_0) \xleftrightarrow{\mathcal{L}} e^{-st_0} \quad (8)$$

*Handwritten notes:*  
 - Above (8): LTI system  
 - Below (8): for pulse response that delays the input by  $t_0$

In words: a system that delays by  $t_0$  has a Laplace transform  $e^{-st_0}$ .



# Example

Find the Laplace transform of

$$x(t) = 3 \cos(4\pi t)u(t) + e^{-3t}u(t-3) \quad (9)$$

We can use linearity and time shifts:

$$x(t) = 3\mathcal{L}\{\cos(4\pi t)u(t)\} + \mathcal{L}\left\{e^{-9}e^{-3(t-3)}u(t-3)\right\} \quad (10)$$

$$= \frac{3s}{s^2 + (4\pi)^2} + \frac{e^{-9}e^{-3s}}{s+3} \quad (11)$$

The ROC is  $\{s : \Re\{s\} > 0\} \cap \{s : \Re\{s\} > -3\} = \{s : \Re\{s\} > 0\}$ .



# $s$ -domain shift

What about a shift in  $s$  instead of a shift in time? Using the same type of argument as time shifts, we get

$$\underline{e^{s_0 t} x(t)} \stackrel{\mathcal{L}}{\longleftrightarrow} \underline{X(s - s_0)} \quad (12)$$

Now the ROC shifts by  $\Re\{s_0\}$ : if  $s_0 > 0$  it shifts to the right, and if  $s_0 < 0$  it shifts to the left.



# Example

Find the Laplace transform of

$$x(t) = e^{-at} \sin(\omega_0 t) u(t). \quad (13)$$

*decaying exp.  
s<sub>0</sub> = -a*

We have

$$\sin(\omega_0 t) u(t) \xleftrightarrow{\mathcal{L}} \frac{\omega_0}{s^2 + \omega_0^2}. \quad (14)$$

*shift this  
(Ls) by -a*

So

$$e^{-at} \sin(\omega_0 t) u(t) \xleftrightarrow{\mathcal{L}} \frac{\omega_0}{(s + a)^2 + \omega_0^2} \quad (15)$$

*where are  
the poles?*

with ROC  $\{\Re\{s\} > -a\}$  (shifted to the left by a).



# Time dilation/compression

If we look at  $x(at)$ , we can plug into the transform to get

$$\underline{x(at)} \stackrel{\mathcal{L}}{\longleftrightarrow} \int_{-\infty}^{\infty} \underline{x(at)} e^{-st} dt = \int_{-\infty}^{\infty} \frac{1}{|a|} x(\tau) e^{-(s/a)\tau} d\tau \quad (16)$$

$$= \frac{1}{|a|} X\left(\frac{s}{a}\right) \quad (17)$$

The ROC is:

$$\mathcal{R}(x(at)) = \{s : s/a \in \mathcal{R}\}. \quad (18)$$

If  $|a| < 1$  we are slowing down  $x(t)$  and the ROC shrinks by a factor of  $a$  and if  $|a| > 1$  we are speeding up  $x(t)$  and the ROC expands by a factor of  $a$ .



# Conjugation

If  $x(t) \xleftrightarrow{\mathcal{L}} X(s)$  then

$$\underbrace{x^*(t)} \xleftrightarrow{\mathcal{L}} = \int_{-\infty}^{\infty} \underbrace{x^*(t)} e^{-st} dt = \left( \int_{-\infty}^{\infty} \underbrace{x(t)} e^{-s^*t} \right)^* \quad (19)$$

$$= X^*(s^*). \quad (20)$$

Why is this useful? If  $x(t)$  is real, then  $x(t) = x^*(t)$  and  $X(s) = X^*(s^*)$ . That means if  $X(s)$  has a pole or zero at  $s = s_0$ , then it also has one at  $s = s_0^*$ .

We say that for real signals, the poles and zeros appear in conjugate pairs.



# Try it yourself

## Problem

*Try finding the following Laplace transforms using the properties:*

$$x(t) = \delta(t - 2) - 3e^{-5t}u(t) \quad (21)$$

$$x(t) = e^{-3t}u(t) + e^{-4t} \cos(6t)u(t) \quad (22)$$

$$x(t) = e^{-2|t|} \quad (23)$$

